# Information, Codes and Ciphers

By Jeremy Le for MATH3411 24T3

# 1 Introduction

# 1.1 Mathematical Model

To give a mathematical framework for digital data transmission, define

- a source alphabet  $S = \{s_1, s_2, \dots, s_q\}$  of q symbols
- a **code alphabet** A of r symbols probabilities  $p_i = P(s_i)$
- a **code** that encodes each symbol  $s_i$  by a codeword which is a **string** of code symbols.

# 1.2 Assumed Knowledge

- Modular Arithmetic and the Division Algorithm
- Probability (Binomial Distribution and Bayes' Rule)
- Linear Algebra (Linear combination, independence, etc...)

#### 1.3 Morse Code

Morse code is a **ternary** code (radix 3). Its alphabet is

- 1.  $\bullet$  called **dot**
- 2. called dash
- 3. p a pause

The codewords are strings of • and — **terminated** by p.

#### 1.4 ASCII

American National Standard Code for Information Interchange.

Binary code of fixed codeword length, namely 7, with  $2^7 = 128$  encoded symbols.

The extended ASCII is a code like the 7-bit ASCII but with an extra bit in the front used as a check bit, requiring the number of 1's to be even.

#### 1.5 ISBN

International Standard Book Number.

They have 10 bits, with it's last bit being a check bit, requiring

$$\sum_{i=1}^{10} ix_i \equiv 0 \pmod{11}.$$

# 2 Error Detection and Correction Codes

We say that  $\mathbf{x}$  corrupted to  $\mathbf{y}$  is denoted by  $\mathbf{x} \rightsquigarrow \mathbf{y}$ .

# 2.1 ISBN-10 Error Capability

ISBN-10 numbers are capable of detecting the two types of errors:

- 1. getting a digit wrong,
- 2. interchanging two (unequal) digits.

# 2.2 Types of Codes

- variable length code: codewords have different lengths
- block code: codewords have the same lengths
- t-error correcting code: code can always correct up to t errors
- systematic code: code with information digits and check digits distinct

# 2.3 Binary Repetition Codes

A binary r-repetition code encodes  $0 \to \overbrace{0 \cdots 0}^r$  and  $1 \to \overbrace{1 \cdots 1}^r$ 

The binary (2t+1)-repetition code is t-error correcting. The binary 2t-repetition code is (t-1)-error correcting and t-error detecting.

# 2.4 Information Rate and Redundancy

The **information rate** R is given by,

- For a code C of radix r and length  $n, R = \frac{\log_r |C|}{n}$
- For a systematic code,  $R = \frac{\text{\# information digits}}{\text{length of code}}$

We then define **redundancy** =  $\frac{1}{R}$ .

# 2.5 Binary Hamming Error-Correcting Codes

A Binary Hamming (n, k) code is a code of length n with k information bits, such that it is a single error correcting and has a parity check matrix, H, of size n - k by n.

# 2.6 Hamming Distance, Weights

The **weight** of an n-bit word  $\mathbf{x}$  is defined to be

$$w(\mathbf{x}) = \#\{i : 1 \le i \le n, x_i \ne 0\}.$$

Given two n-bit words, the **Hamming distance** between them is

$$d(\mathbf{x}, \mathbf{y}) = \#\{i : 1 \le i \le n, x_i \ne y_i\}.$$

Given some code with set of codewords C, we define (minimum) weight of C to

$$w = w(C) = \min\{w(\mathbf{x}) : \mathbf{x} \in C, \mathbf{x} \neq \mathbf{0}\}.$$

Similarly, the (minimum) distance of C is defined by

$$d = d(C) = \min\{d(\mathbf{x}, \mathbf{y}) : \mathbf{x}, \mathbf{y} \in C, \mathbf{x} \neq \mathbf{y}\}.$$

If  $\mathbf{x} \leadsto \mathbf{y}$ , then  $d(\mathbf{x}, \mathbf{y})$  is the number of errors in  $\mathbf{y}$ .

# 2.7 Decoding Strategies

Minimum Distance Decoding Strategy Given a received word y, decode to *closest* codeword x.

**Standard Strategy** If received word y is distance at most t from a codeword x, then decode y to x; otherwise flag an error.

**Pure Error Detection** If received word y is not a codeword x, then flag an error.

# 2.8 Sphere Packing

The **sphere** of radius r around c:

$$S_r(\mathbf{c}) = \{ \mathbf{x} \in \mathbb{Z}_2^n : d(\mathbf{x}, \mathbf{c}) \le r \}.$$

The volume of this sphere is its size  $|S_r(\mathbf{c})|$ .

**Sphere-Packing Condition Theorem** A t-error correcting binary code C of length n has minimum distance d = 2t + 1 or 2t + 2, and

$$|C|\sum_{i=0}^{t} \binom{n}{i} \le 2^n.$$

If we have equality in the bound, then we say that the code is perfect. This means that codewords are evenly spread around in  $\mathbb{Z}_2^n$  space.

More generally for radiux r:

$$|C| \le \frac{r^n}{\sum_{i=0}^t \binom{n}{i} (r-1)^i}.$$

## 2.9 Binary Linear Codes

A linear code C is a vector space over some field  $\mathbb{F}$ . Equivalently it is the null-space of

$$C = \{ \mathbf{x} \in \mathbb{F}^n : H\mathbf{x}^T = \mathbf{0} \}$$

of an  $m \times n$  parity check matrix H with m = rank(H).

- $\dim C = k = n m$  by the Rank-Nullity Theorem.
- If C is binary, then  $|C| = 2^k$ .
- C is systematic.
- If *H* is **reduced echelon form**, then we can choose the non-leading columns of *H* to be **information bits** and the leading columns of *H* to be **check bits**.

Minimum Distance for Linear Codes If C is a linear code with parity check matrix H, then

- w(C) = d(C),
- $d(C) = \min\{r : H \text{ has } r \text{ linearly dependent columns}\}.$

For a linear code C, the **row space** (or **row span**) of a  $k \times n$  **generator matrix** G over  $\mathbb{F}$  generates C, in the sense that C is a set of linear combinations of G.

#### 2.10 Standard Form Matrices

For a linear code C of dimension k and length n = k + m,

- $H = (I_m \mid B)$  is a parity check matrix for C if and only if
- $G = (-B^T \mid I_k)$  is a generator matrix C.

Linear codes C and C' are **equivalent** if C' is obtained by permuting the codeword entries of C by a fixed permutation:

$$C' = CP = \{ \mathbf{x}P : \mathbf{x} \in C \}$$
 for some permutation matrix  $P$ 

Note that G' = GP and H' = HP.

## 2.11 Extending Linear Codes

The **extension** of a linear code C:

$$\hat{C} = \{x_0 x_1 \cdots x_n : x_1 \cdots x_n \in C, x_0 = -(x_1 + \cdots + x_n)\}.$$

The **extension**  $\hat{C}$  is a linear code with minimum distance d(C) or d(C) + 1.

#### 2.12 Radix r Hamming Codes

- Let r be a prime number and  $m \ge 1$  some integer.
- Write the numbers  $1, \ldots, r^m 1$  in base r, as length m column vectors.
- Of each set of r-1 parallel columns, delete all whose first nonzero entry is not 1.
- This gives the radix r Hamming code of length  $n = \frac{r^m 1}{r 1}$ .

# 3 Compression Coding

#### **Definitions**

source $S$	with	symbols	$s_1, \ldots, s_q$
	with	probabilities	$p_1,\ldots,p_q$
$\operatorname{\mathbf{code}} C$	with	codewords	$\mathbf{c}_1,\dots,\mathbf{c}_q$
		of lengths	$\ell_1,\ldots,\ell_q$
		and $\mathbf{radix}$	r

#### 3.1 Instantaneous and UD Codes

A code C is

- uniquely decodable (UD) if it can always be decoded unambiguously
- **instantaneous** if no codeword is a **prefix** of another. Such a code is an **I-code**.

Decision trees can represent I-codes.

- Branches are numbered from the top down.
- Any radix r is allowed.
- Two codes are equivalent if their decision trees are isomorphic.
- By shuffling source symbols, we may assume that  $\ell_1 \leq \ell_2 \leq \cdots \leq \ell_q$ .

The Kraft-Mcmillan Theorem The following are equivalent:

- 1. There is a radix r **UD-code** with codeword lengths  $\ell_1 \leq \ell_2 \leq \cdots \leq \ell_q$
- 2. There is a radix r **I-code** with codeword lengths  $\ell_1 \leq \ell_2 \leq \cdots \leq \ell_q$
- 3.  $K = \sum_{i=1}^{q} (\frac{1}{r})^{\ell_i} \le i$

#### 3.2 Minimal UD-Codes

The (expected or) average length and variance of codewords in C are

$$L = \sum_{i=1}^{q} p_i \ell_i$$
  $V = (\sum_{i=1}^{q} p_i \ell_i^2) - L^2$ 

A UD-code is **minimal** with respect to  $p_1, \ldots, p_q$  if it has minimal length.

**Minimal UD-Codes** If a UD-code has minimal average length L with respect to  $p_1, \ldots, p_q$ , then, possibly after permuting codewords of equally likely symbols,

- $\ell_1 \leq \ell_2 \cdots \leq \ell_q$
- $\ell_{q-1} = \ell_q$
- If C is instantaneous, then  $\mathbf{c}_{q-1}$  and  $\mathbf{c}_q$  differ only in their last place.
- $\bullet$  If C is binary, then

$$K = \sum_{i=1}^{q} 2^{-\ell i} = 1$$

## 3.3 Huffman's Algorithm

#### **Binary Case**

- 1. Write the symbols in a column, with highest probability at the top and lowest probability at the bottom.
- 2. Merge the bottom two (least frequent) symbols  $s_q$  and  $s_{q-1}$  into one big symbol of probability  $p_q + p_{q-1}$ .
- 3. Write the resulting q-1 symbols in a new column to the right in same order as before. Make sure to place the newly created symbol as high as possible in this column.
- 4. Draw branches from the newly created symbol to its two constituent symbols, and label them 0 and 1.
- 5. Repeat the above, until there is only one symbol left.

**Huffman Code Theorem** For any given source S and corresponding probabilities, the Huffman Algorithm yields an instantaneous minimum UD-code.

**Knuth** The average codeword length L of each Huffman code is the sum of all child node probabilities.

#### 3.4 Extensions

For a source  $S = \{s_1, \ldots, s_q\}$  with probabilities  $p_1, \ldots, p_q$ , the n-th extension of S is the Cartesian product  $S^n$ , containing all strings of n symbols in S.

The probability of each symbol in  $S^n$  is the product of the probabilities of constituent symbols. We also order the new symbols in non-increasing probability.

#### 3.5 Markov Sources

A k-memory source S is one whose symbols each depend on the previous k.

- If k = 0, then no symbol depends on any other, and S is memoryless.
- If k = 1, then S is a Markov source.
- $p_{ij} = P(s_i \mid s_j)$  is the probability of  $s_i$  occurring right after a given  $s_j$ .
- The matrix  $M = (p_{ij})$  is the **transition matrix**.
- Entry  $p_{ij}$  is the probability of getting from state  $s_j$  to state  $s_i$ .

A Markov process M is in equilibrium  $\mathbf{p}$  if  $\mathbf{p} = M\mathbf{p}$ .

We will assume that

- M is **ergodic**: we can get from any state j to any state i.
- M is aperiodic: the gcd of cycle lengths is 1.

Under the above assumptions, M has a non-zero equilibrium state.

### 3.6 Arithmetic Coding

Consider a source  $\{s_1, \ldots, s_q\}$  where  $s_q = \bullet$  is called a stop symbol, with probabilities  $p_1, \ldots, p_q$ . In this context, a message will always end with a stop symbol. Encoding a message  $s_{i_1} \ldots s_{i_n}$  involves the following steps:

- Split up the interval [0,1) into sub-intervals of size  $p_1,\ldots,p_q$ .
- Choose the  $i_1$ -th sub-interval.
- Split up this sub-interval again, in proportion to  $p_1, \ldots, p_q$ .
- Choose the  $i_2$ -th sub-interval.
- Repeat this for the rest of the symbols, and output any number inside the final sub-interval found.

#### 3.7 Dictionary Methods

**Encoding** Consider a message  $m = m_1 m_2 \dots m_n$ . To encode m:

- Begin with an empty table D, and set the 0-th entry to  $\emptyset$ , representing an empty string.
- Find the longest prefix s of m in D (possibly the empty string  $\emptyset$ ), and say s is in entry k.
- Find the symbol c just after s.
- Add a new entry sc to D, remove sc from m, and output (k, c).
- Repeat until m is fully encoded.

**Decoding** Consider an encoded message  $(k_1, c_1) \dots (k_n, c_n)$ . To decode this message, take the following steps:

- Begin with a table D with  $\emptyset$  in the 0-th entry.
- Let  $s_1$  be the  $k_1$ -th entry in the table. Append  $s_1c_1$  to the table, and output  $s_1c_1$ .
- Let  $s_2$  be the  $k_2$ -th entry in the table. Append  $s_2c_2$  to the table, and output  $s_2c_2$ .
- Keep doing this until the message is fully decoded.

# 4 Information Theory

Define  $I(s_i) = I(p_i) = -\log_2 p_i$ . Define the (Shannon) **entropy** of S:

$$H_r(S) = \sum_{i=1}^{q} p_i I_r(p_i) = -\sum_{i=1}^{q} p_i \log_r p_i$$

This expresses the average information per source symbol.

**Gibb's Inequality** If  $p_1, \ldots, p_q$  and  $p'_1, \ldots, p'_q$  are probability distributions, then

$$-\sum_{i=1}^{q} p_{i} \log_{r} p_{r} \le -\sum_{i=1}^{q} p_{i} \log_{r} p'_{i}.$$

Equivalently,

$$\sum_{i=1}^{q} p_i \log_r \frac{p_i'}{p_i} \le 0.$$

Furthermore, there is equality if and only if  $p_i = p'_i$  for all i.

**Maximum Entropy Theorem** For any source S with q symbols, the base r entropy satisfies

$$H_r(S) \le \log_r q$$

with equality if and only if all symbols are equally likely.

First Source-Coding Theorem For each radix r UD-code C for source S,

$$H_r(S) < L_r$$

with equality iff  $p_i = r^{-\ell_i}$  for all i and  $K_r = \sum_{i=1}^q r^{-\ell_i} = 1$ .

# 4.1 Entropy of Extensions for Memoryless Sources

**Entropy of Extensions** 

$$H_r(S^n) = nH_r(S).$$

Sannon's Source Coding Theorem Encoding  $S^n$  by an SF-code or a Huffman code allows the average codeword lengths to be arbitrarily close to the entropy:

$$\frac{L_r^{(n)}}{n} \to H_r(S)$$
 for  $n \to \infty$ .

#### 4.2 Entropy for Markov Sources

Consider a Markov source  $S = \{s_1, \ldots, s_q\}$  with probabilities  $p_1, \ldots, p_q$ , transition matrix  $M = (p_{ij}) = (P(s_i \mid s_j))$  and equilibrium  $\mathbf{p} = (p_i)$ .

The **conditional information** of  $s_i$  given  $s_j$  is

$$I(s_i \mid s_j) = -\log P(s_i \mid s_j) = -\log p_{ij}.$$

The conditional entropy given  $s_i$  is

$$H(S \mid s_j) = \sum_{i=1}^{q} p_{ij} I(s_i \mid s_j) = -\sum_{i=1}^{q} P(s_i \mid s_j) \log P(s_i \mid s_j)$$

The Markov entropy of S is

$$H_M(S) = \sum_{j=1}^{q} p_j H(S \mid s_j)$$

$$= -\sum_{i=1}^{q} \sum_{j=1}^{q} p_j p_{ij} \log p_{ij}$$

$$= -\sum_{i=1}^{q} \sum_{j=1}^{q} P(s_j s_i) \log P(s_i \mid s_j).$$

The equilibrium entropy of S is

$$H_E(S) = -\sum_{j=1}^q p_j \log p_j.$$

**Theorem on Markov Entropy** For a Markov source S,

$$H_M(S) \leq H_E(S)$$
.

There is equality if and only if the symbols in S are independent.

# 4.3 Noisy Channels

Source entropy:  $H(A) = -\sum_{j=1}^{n} P(a_j) \log P(a_j)$ 

Output entropy:  $H(B) = -\sum_{i=1}^{v} P(b_i) \log P(b_i)$ 

Conditional entropies:  $H(B \mid a_j) = -\sum_{i=1}^{v} P(b_i \mid a_j) \log P(b_i \mid a_j)$ 

 $H(A \mid b_i) = -\sum_{j=1}^{u} P(a_j \mid b_i) \log P(a_j \mid b_i)$ 

Joint entropy:  $H(A,B) = -\sum_{i=1}^{v} \sum_{j=1}^{u} P(a_j \cap b_i) \log P(a_j \cap b_i)$ 

# 4.4 Channel Capacity

The channel capacity is

$$C = C(A, B) = \max I(A, B)$$

where the maximum is taken over all possible probabilities for A's symbols.

**Theorem** The channel capacity of a binary symmetric channel with crossover probability p is 1 - H(p).

# 5 Number Theory and Algebra

#### 5.1 Revision of Discrete Mathematics

- Division Algorithm
- Inverses
- (Extended) Euclidean Algorithm
- Chinese Remainder Theorem
- Bezout's Identity

# 5.2 Number Theory Results

Given  $m \in \mathbb{Z}^+$ , the set of invertible elements in  $\mathbb{Z}_m$  is denoted by

$$\mathbb{U}_m = \{ a \in \mathbb{Z}_m : \gcd(a, m) = 1 \}$$

and its elements are the **units** in  $\mathbb{Z}_m$ . Euler's **phi-function** is defined by

$$\phi(m) = |\mathbb{U}_m|.$$

Formula for  $\phi(m)$ 

- 1. If gcd(m, n) = 1, then  $\phi(mn) = \phi(m)\phi(n)$ .
- 2. For a prime p and  $\alpha \in \mathbb{Z}^+$ , we have  $\phi(p^{\alpha}) = p^{\alpha} p^{\alpha-1}$ .
- 3. Hence, if  $m = p_1^{\alpha_1} p_2^{\alpha_2} \cdots p_r^{\alpha_r}$  is the prime factorisation of m, then

$$\phi(m) = (p_1^{\alpha_1} - p_1^{\alpha_1 - 1})(p_2^{\alpha_2} - p_2^{\alpha_2 - 1}) \cdots (p_r^{\alpha_r} - p_r^{\alpha_r - 1}).$$

**Primitive Element Theorem** Given a prime p, there exists  $g \in \mathbb{U}_p$  such that

$$\mathbb{U}_p = \{g^0 = 1, g, g^2, \dots, g^{p-2} \text{ and } g^{p-1} = 1.$$

**Primitve Powers** If g is primitive in  $\mathbb{Z}_p$ , then  $g^k$  is primitive if and only if  $\gcd(k, p-1) = 1$  and hence there are  $\phi(p-1)$  primitive elements in  $\mathbb{Z}_p$ .

**Euler's Theorem** If gcd(a, m) = 1, then  $a^{\phi(m)} \equiv 1 \pmod{m}$ .

**Corollary** If gcd(a, m) = 1, then  $ord_m(a) \mid \phi(m)$ .

**Fermat's Little Theorem** For prime p and any  $a \in \mathbb{Z}$ ,  $a^p \equiv a \pmod{p}$ 

#### 5.3 Finite Fields

Finite Field Theorem If p is prime, m(x) a monic irreducible in  $\mathbb{Z}_p[x]$  of degree n, and  $\alpha$  denotes a root of m(x) = 0, then

- 1.  $\mathbb{F} = \mathbb{Z}_p[x]/\langle m(x) \rangle$  is a field,
- 2.  $\mathbb{F}$  is a vector space of dimension n over  $\mathbb{Z}_p$ ,
- 3.  $\mathbb{F}$  has  $p^n$  elements,
- 4.  $\{\alpha^{n-1}, \alpha^{n-2}, \dots, \alpha, 1\}$  is a basis for  $\mathbb{F}$ ,
- 5.  $\mathbb{F} = \mathbb{Z}_p(\alpha)$  i.e. the smallest field containing  $\mathbb{Z}_p$  and  $\alpha$ ,
- 6. there exists a primitive element  $\gamma$  of order  $p^n-1$  for which  $\mathbb{F}=\{0,1,\gamma,\gamma^2,\ldots,\gamma^{p^{n-2}}\},$

if a field  $\mathbb{F}$  has a finite number of elements, then  $|\mathbb{F}| = p^n$  where p is prime, and  $\mathbb{F}$ is isomorphic to  $\mathbb{Z}_p[x]/\langle,(x)\rangle$ . Hence ALL fields with  $p^n$  elements are isomorphic to one another.

## 5.4 Primality Testing

**Pseudo-Prime Test** No if we fine n is composite.

- Let  $a \in \mathbb{N}$  with a < n.
  - If  $gcd(a, n) \neq 1$ , then n must be composite so return no
  - Otherwise, if  $a^{n-1} \neq 1 \pmod{n}$ , then n is composite so return no

**Lucas' Test** Possible answer as to whether n is prime

- Let  $a \in \mathbb{N}$  with a < n.
  - If  $gcd(a, n) \neq 1$ , then n must be composite so no
  - If  $a^{n-1} \neq 1 \pmod{n}$ , then n must be composite so no
  - If  $a^{(n-1)/p} \neq 1 \pmod{n}$  for all primes  $p \mid n-1$ , then ves

Miller-Rabin Test If n is composite, then return no

- Write  $n = 2^s t + 1$  where t is odd
- Choose some  $a \in \{1, ..., n-1\}$  at random
- If  $a^t \equiv 1 \pmod{n}$ , then it is probably prime
- For r = 0, ..., s 1,
  - If  $a^{2^r t} \equiv -1 \pmod{n}$ , then n is probably prime.
- Otherwise, it is not prime.

**Fermat Factorisation** a two-factorisation of n

- For  $t = \lceil \sqrt{n} \rceil, \dots, n$ :
  - If  $s^2 = t^2 n$  is square, then return n = ab = (t-s)(t+s)

# 6 Algebraic Coding

# 6.1 Single Error Correcting BCH Codes

Let  $f(x) \in \mathbb{Z}_2[x]$  be a polynomial of degree m with a primitive root  $\alpha$ . Let  $n = 2^m - 1$  and k = n - m.

The matrix  $H = \begin{pmatrix} 1 & \alpha & \alpha^2 & \dots & \alpha^{n-1} \end{pmatrix}$  is the check matrix of a binary Hamming (n, k) code C. Every binary Hamming (n, k) code can be obtained in this way.

Let  $\mathbf{c} = (c_0, c_1, \dots, c_{n-1}) \in C$  be a codeword.

- $c_0, \ldots, c_{m-1}$  are the check bits
- $c_m, \ldots, c_{n-1}$  are the information bits

The first m bits of the codeword are check bits, since the first m columns of H are the leading columns. Therefore the information bits and the check bits are neatly divided.

The syndrome of the codeword  $\mathbf{c}$  is

$$S(\mathbf{c}) = H\mathbf{c}^T = c_0 + c_1\alpha + c_2\alpha^2 + \dots + c_{n-1}\alpha^{n-1} = C(\alpha)$$

where  $C(x) = c_0 + c_1 x + c_2 x^2 + \cdots + c_{n-1} x^{n-1}$  is the codeword polynomial corresponding to **c**.

- Since **c** is the codeword, its syndrome is 0. Therefore,  $S(\mathbf{c}) = 0 = C(\alpha)$ , so  $\alpha$  is a root of C(x).
- Since  $\alpha$  is a root of C(x), the minimal polynomial,  $M_1(x)$  of  $\alpha$  must divide C(x) without a remainder.
- $M_1(x)$  is the primitive polynomial f(x).

#### **BCH Encoding**

1. From our message  $c_m, \ldots, c_{n-1}$ , we form the **information** polynomial

$$I(x) = c_m x^m + c_{m+1} x^{m+1} + \dots + c_{n-1} x^{n-1}$$

2. Using polynomial long division, we find the **check polynomial** of degree at most m-1

$$R(x) = I(x)(\mod M_1(x)) = c_0 + c_1 x + \dots + c_{m-1} x^{m-1}.$$

3. Calculate the codeword polynomial

$$C(x) = I(x) + R(x).$$

The codeword is  $(c_0, c_1, \ldots, c_{n-1})$  where the first m bits are check bits and the last k bits are information bits.

BCH Error Correcting and Decoding Suppose that we receive  $\mathbf{d} = \mathbf{c} + e_j$ , where  $e_j$  is a unit vector with 1 in the  $a^j$  position and zero entries elsewhere.

- 1. Let us represent  ${\bf c}$  and  ${\bf d}$  as codeword polynomials C(x) and D(x)
- 2. Calculating the syndrome of d gives us

$$S(\mathbf{d}) = D(\alpha) = C(\alpha) + a^j = a^j$$

The error is therefore in the  $\alpha^j$  position, which is the  $j+1^{th}$  letter of the code. If the syndrome of the codeword we receive is 0, i.e  $D(\alpha) = 0$ , then there is no error.

3. To decode a codeword, look at the last k bits, which are the information bits  $(c_m, \ldots, c_{n-1})$ .

# 6.2 Two Error Correcting BCH Codes

**Theorem** If p is prime and  $\beta$  is a root of  $f(x) \in \mathbb{Z}_2[x]$  then so is  $\beta^{p^i}$  for all i.

#### Construction

- 1. Find a primitive root of minimal polynomial  $M_1(x)$ , with cyclotomic coset  $K_1$
- 2. Select an index not belonging to  $K_1(i \in \{1, ..., p^m 1\} \setminus K_1)$
- 3. Find the minimal polynomial  $M_i(x)$  for  $\alpha^i$
- 4. Define  $M(x) = M_1(x)M_i(x)$
- 5. Define the check matrix

$$H = \begin{pmatrix} 1 & \alpha & \dots & \alpha^{n-1} \\ 1 & \alpha^i & \dots & (\alpha^i)^{n-1} \end{pmatrix}$$

6. Define the syndrome of a codeword  $\mathbf{c} = (c_0, c_1, \dots, c_{n-1})$  as

$$S(\mathbf{c}) = H\mathbf{c}^T = \begin{pmatrix} c_0 + c_1\alpha + \dots + c_{n-1}\alpha^{n-1} \\ c_0 + c_1\alpha^i + \dots + c_{n-1}(\alpha^i)^{n-1} \end{pmatrix} = \begin{pmatrix} C(\alpha) \\ C(\alpha^i) \end{pmatrix}$$

**Encoding and Decoding** Encoding a double-error correcting BCH code is the same as encoding a single-error correcting BCH code, with the difference being

$$R(x) = I(x) \pmod{M(x)}$$

where M(x) is used instead of  $M_1(x)$ .

Error Correcting and Decoding Suppose that we received  $\mathbf{d} = \mathbf{c} + e_i + e_l$ .

- 1. Let us represent  ${\bf c}$  and  ${\bf d}$  as codeword polynomials C(x) and D(x)
- 2. Calculate the syndrome

$$S(\mathbf{c}) = \begin{pmatrix} D(\alpha) \\ D(\alpha^i) \end{pmatrix} = \begin{pmatrix} C(\alpha) + a^j + a^l \\ C(\alpha^i) + (\alpha^i)^j + (\alpha^i)^l \end{pmatrix} = \begin{pmatrix} \alpha^j + \alpha^l \\ \alpha^{ij} + \alpha^{il} \end{pmatrix} = \begin{pmatrix} C(\alpha) + a^j + a^l \\ C(\alpha^i) + (\alpha^i)^j + (\alpha^i)^l \end{pmatrix}$$

The syndrome allows us to determine when there is 0, 1 or 2 errors.

- 0 errors when  $D(\alpha) = D(\alpha^i) = 0$
- 1 error when  $D(\alpha) \neq 0$  and  $D(\alpha)^i = D(\alpha^i)$
- 2 errors when  $D(\alpha) \neq 0$  and  $D(\alpha)^i \neq D(\alpha^i)$

# 7 Cryptography (Ciphers)

# 7.1 Some Classical Cryptosystems

- Caesar Ciphers
- Simple (monoalphabetic) substitution Cipher
- Transposition Cipher
- Combined Systems
- Polyalphabetic Substitution Ciphers
- Non-Periodic Polyalphabetic Substitutions

**Kasiski's Method** For a message m of length n, the index of coincidence is given by

$$I_c = \frac{\sum (f_i^2) - n}{n^2 - n}$$

where  $f_i$  is the frequency of each letter in the message. Solving for r we get

$$r \approx \frac{0.0273n}{(n-1)I_c - 0.0385n + 0.0658}$$

## 7.2 Unicity Distance

The unicity distance is

$$n_0 = \lceil \frac{H_2(K)}{\log_2 q - R} \rceil$$

where K is the total number of keys, q is the number of letters in the source alphabet, and R is the rate of the language in bits per character.

For English text, q = 26 and  $R \approx 1.5$ . So

$$n_0 = \lceil \frac{H_2(K)}{\log_2 q - R} \rceil \approx \lceil \frac{H_2(K)}{4.7 - 1.5} \rceil = \lceil \frac{H_2(K)}{3.2} \rceil.$$

If the keys are equally likely, then

$$n_0 \approx \lceil \frac{log_2|K|}{3 \ 2} \rceil$$